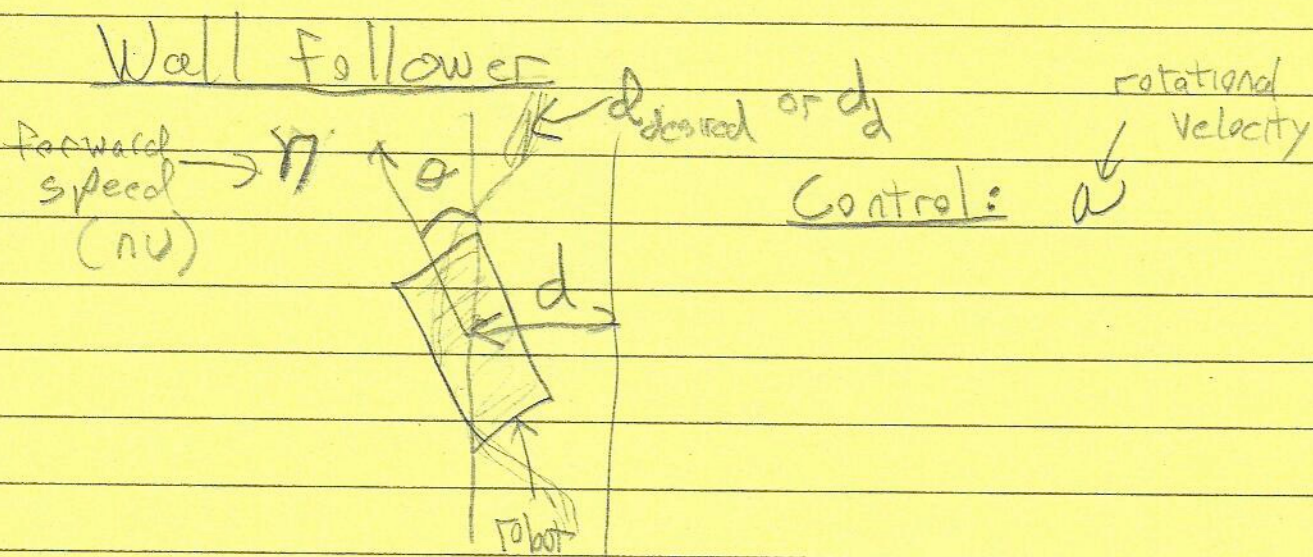


6.3100

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①

I_n D.Tlinearized sin(θ)I_n C.T.

$$d[n] = d[n-1] + \Delta T v \theta[n-1]$$

$$\theta[n] = \theta[n-1] + \Delta T \gamma c[n-1]$$

$c[n] = d_d - d[n]$ control

$$c[n] = K_p (d_d[n] - d[n])$$

proportional control

$$\frac{d[n] - d[n-1]}{\Delta T}$$

$$\frac{d}{dt} d(t) = v \theta(t)$$

$$\frac{d}{dt} \theta(t) = \gamma c(t)$$

$$c(t) = K_p (d_d(t) - d(t))$$

$$\begin{bmatrix} d[n] \\ \theta[n] \end{bmatrix} = \underbrace{\begin{bmatrix} 1 & \Delta T v \\ \Delta T \gamma K_p & 1 \end{bmatrix}}_{A_{dt}} \begin{bmatrix} d[n-1] \\ \theta[n-1] \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ \Delta T \gamma K_p \end{bmatrix}}_{B_{dt}} d_d[n]$$

$$\frac{d}{dt} \begin{bmatrix} d(t) \\ \theta(t) \end{bmatrix} = \underbrace{\begin{bmatrix} 0 & v \\ \gamma K_p & 0 \end{bmatrix}}_{A_{ct}} \begin{bmatrix} d(t) \\ \theta(t) \end{bmatrix} + \underbrace{\begin{bmatrix} 0 \\ \gamma K_p \end{bmatrix}}_{B_{ct}} d_d(t)$$

D.T.

$$x[n] = A x[n-1] + B u[n-1]$$

$$x[n] = \underbrace{A^n x[0]}_{\text{ZIR}} + \underbrace{\sum_{m=0}^{n-1} A^{n-m-1} B u[m]}_{\text{ZSR}}$$

if $|\lambda(A)| < 1 \forall i$ then ZIR $\rightarrow 0$ as $n \rightarrow \infty$

C.T.

$$\frac{d}{dt} x(t) = A x(t) + B u(t)$$

$$x(t) = \underbrace{e^{At} x(0)}_{\text{ZIR}} + \underbrace{\int_0^t e^{A(t-\tau)} B u(\tau) d\tau}_{\text{ZSR}}$$

What's e^{At}

etc $\downarrow$ eval's

$$\text{If } A = V \begin{bmatrix} \lambda_1 & & \\ & \ddots & \\ & & \lambda_n \end{bmatrix} V^{-1}$$

$$e^{At} = V \begin{bmatrix} e^{\lambda_1 t} & & \\ & \ddots & \\ & & e^{\lambda_n t} \end{bmatrix} V^{-1}$$

$$\text{If } \operatorname{Re}(\lambda_i(A)) < 0 \forall i$$

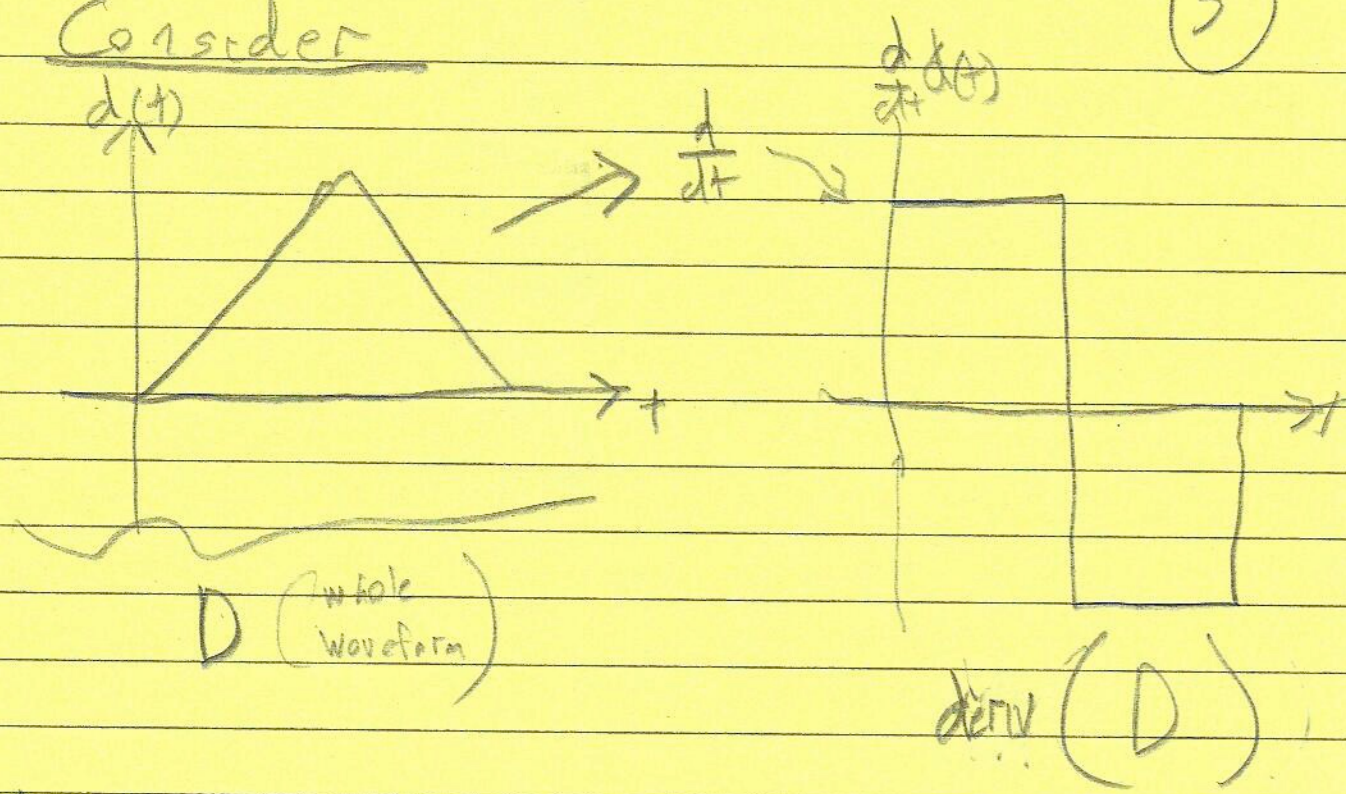
then ZIR $\rightarrow 0$ as $t \rightarrow \infty$

ZIR is weighted combo

Wait: What about blocks, and intuition and insight and...

3

Consider



$\frac{d}{dt}$ is an operator, mapping D to its derivative

Use 's' to denote this operator

$$\text{So } \frac{d}{dt} d(t) = \gamma d(t) \Rightarrow s D = \gamma \underline{\Phi}$$

$$\frac{d}{dt} \theta(t) = \gamma \theta(t) \Rightarrow s \underline{\Phi} = \gamma \underline{\Gamma}$$

Some identities

$$s^2 \underline{\Gamma} = s(s \underline{\Gamma}) \Rightarrow \frac{d}{dt} \left(\frac{d}{dt} \gamma(t) \right)$$

$$(Bs + \alpha) \underline{\Gamma} \Rightarrow B \frac{d}{dt} \gamma(t) + \alpha \gamma(t) = \frac{d^2}{dt^2} \gamma(t)$$

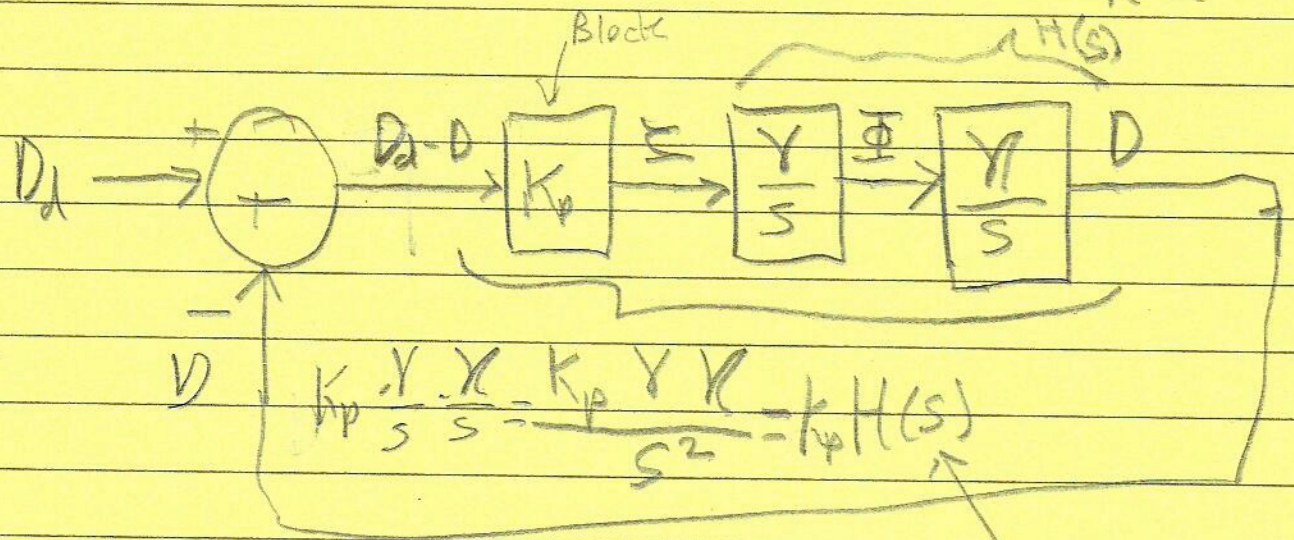
$$(s + \alpha_1)(s + \alpha_2) \underline{\Gamma} = (s^2 + (\alpha_1 + \alpha_2)s + \alpha_1 \alpha_2) \underline{\Gamma} \Rightarrow \frac{d^2}{dt^2} \gamma(t) + (\alpha_1 + \alpha_2) \frac{d}{dt} \gamma(t) + \alpha_1 \alpha_2 \gamma(t)$$

Back to Line Follower

(4)

$$\begin{aligned} s D &= \eta \Phi \\ s \Phi &= \gamma \underline{C} \\ \underline{C} &= K_p (D_d - D) \end{aligned}$$

$$\begin{aligned} D &= \frac{\eta}{s} \Phi \\ \Phi &= \frac{\gamma}{s} \underline{C} \\ \underline{C} &= K_p (D_d - D) \end{aligned}$$



Transfer Operator

$$D = \frac{K_p \gamma \eta}{s^2} (D_d - D)$$

$$\left(1 + \frac{K_p \gamma \eta}{s^2}\right) D = \frac{K_p \gamma \eta}{s^2} D_d$$

$$D = \frac{\frac{K_p \gamma \eta}{s^2}}{1 + \frac{K_p \gamma \eta}{s^2}} D_d$$

$$D = \frac{K_p \gamma \eta}{s^2 + K_p \gamma \eta} D_d$$

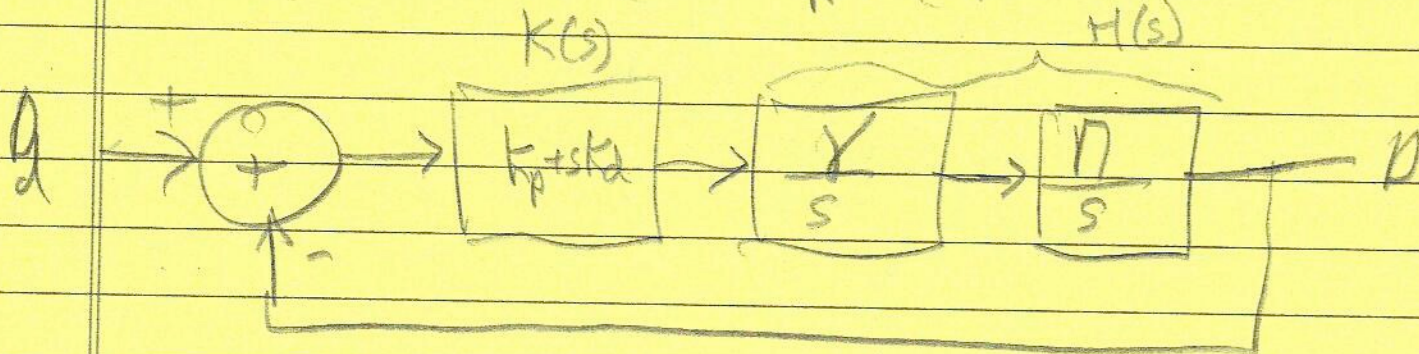
$$(s^2 + K_p \gamma \eta) D = K_p \gamma \eta D_d$$

P. P. controller

(6)

$$c(t) = K_p(d_d(t) - d(t)) + K_d s(d_d(t) - d(t))$$

$$C = (K_p + sK_d) \underbrace{(d_d - d)}_{H(s)}$$



$$K(s)H(s) = (K_p + sK_d) \left(\frac{Y}{s} \cdot \frac{N}{s} \right)$$

$$D = \frac{K(s)H(s)}{1 + K(s)H(s)} D_d$$

$$= \frac{(K_p + sK_d)(YN)}{s^2 + (K_p + sK_d)YN}$$

Find roots

$$(s^2 + sK_dYN + K_pYN) D = (K_p + sK_d) D_d$$

Substitute

$$s^2 + sK_dYN + K_pYN = 0$$

$$\frac{1}{1P} \omega \in \mathbb{R}$$

roots are natural frequencies