

04/16/25.

\* Optimize the tracking in state feedback:  $K_r$  and Integral.

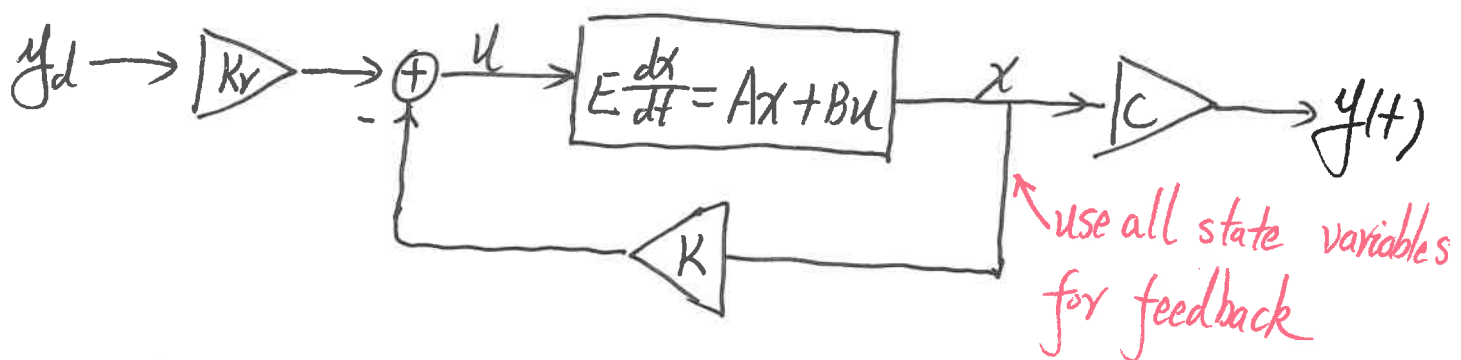
1. Review of State Feedback.

State space model:  $E \frac{dx}{dt} = Ax + Bu$

output:  $y = Cx$

control:  $u = K_r y_d - Kx$ .

Block Diagram:



Natural Response: stability.

close-loop state space model: substitute  $u(t)$ .

$$\frac{d}{dt}x = (A - BK)x + BK_r y_d. \quad (\text{assume } E = I, \text{ identity})$$

Natural frequency:  $\lambda = \text{eig}(A - BK)$ , i.e. solve  $|sI - (A - BK)| = 0$ .

Want all the  $\lambda$ 's to be on the left half plane.

Model system: Given  $K = [K_1, K_2 \dots K_n] \implies$  calculate  $\lambda_{1,2,3 \dots}$  <sup>2</sup>

Design controller: find corresponding  $K$ .  $\Leftarrow$  choose  $\lambda_{1,2,3 \dots}$

usually get a  $n$ th order characteristic equation from  $|sI - (A - BK)| = 0$ .

$(s - \lambda_1)(s - \lambda_2) \dots (s - \lambda_n) = 0 \implies$  expand to get  $K_{1,2 \dots n}$ .

Using matlab / python, simply call function:

$$K = \text{place}(A, B, [\lambda_1, \lambda_2 \dots])$$

define system    specify pole location.

See Propeller Arm Example in slides.

In the 1st example we see the system converges fast to a stable state, however, the final value of  $\theta$  is not the same as  $\theta_d(t=\infty) = 1$ .

2. Steady State of step response.

$$\frac{d}{dt}x = (A - BK)x + BK_r y_d$$

Recall in D.T. for  $n \rightarrow \infty$ ,  $x[n-1] = x[n] = x[\infty]$ .

For C.T.:  $x(\infty) = \text{const}$ , so  $\frac{d}{dt}x(\infty) = 0$ .

$$0 = (A - BK)x(\infty) + BK_r y_d$$

$$x(\infty) = -(A - BK)^{-1} BK_r y_d \quad (\text{note order of matrix multiplication!})$$

Output:  $y(\infty) = Cx(\infty) = -C(A-BK)^{-1}BK_r y_{d0}$  (scalar) → want to be 1 3.

For good tracking performance, want  $y(\infty) = y_{d0}$ .

$$K_r = - \frac{1}{C(A-BK)^{-1}B}$$

See example 2 & 3 from slides where  $K_r$  is fixed.

Last lecture, also discussed another way to determine  $K$ :

Linear Quadratic Regulator:

Ideally want all  $\lambda$ 's to be very negative, however limited by actuator.

Define cost function:  $Z = \int_0^\infty \underbrace{\sum_i q_i x_i^2(t)}_{\text{penalty on bad tracking}} + \underbrace{\gamma u^2(t)}_{\text{penalty on actuation}} dt$ .

Would like to  $\min_{K's} Z(K)$  Have the freedom to assign weight factor.

Matlab:  $K = \text{lqr}(A, B, Q, R)$ .

See example 4 from slides.

3. Adding Integral term as an additional state:

In D.T. (Lab 2 B), we learned using PID control to improve tracking.

To model, we defined  $\text{sum}[n] = \text{sum}[n-1] + \text{err}[n-1] \cdot \Delta T$ .

Using Integral for feedback is advantageous as  $K_y$  relies on  $A, B$ , 4  
i.e. modelling accuracy.

Define a new state:  $J(t) = \int_0^t (y(\tau) - y_d(\tau)) d\tau$ .

$$\text{or } \frac{d}{dt} J(t) = y(t) - y_d(t)$$

Add as the 1st state in the state space model:

$$\frac{d}{dt} J(t) = \underbrace{C X(t)}_{\substack{\downarrow \\ y(t)}} - y_d \downarrow 1$$

$$\frac{d}{dt} X(t) = A X + B U \downarrow N$$

See slides for Example 5:

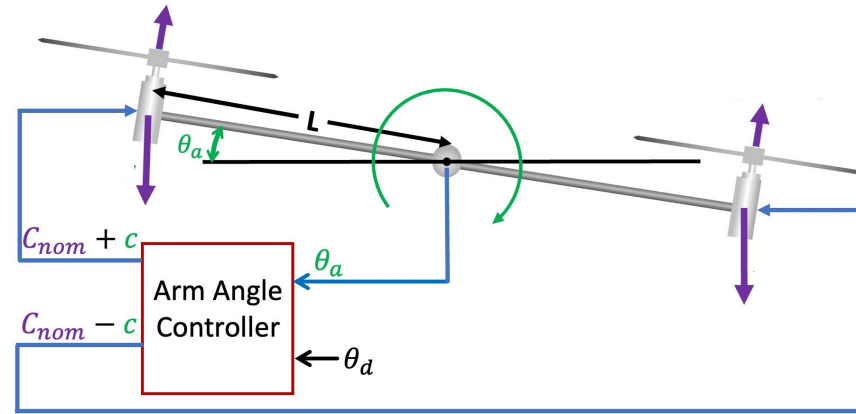
Define ~~an~~ "Augmented" matrices:

$$A_+ = \begin{pmatrix} 0 & C \\ 0 & A \end{pmatrix} \quad B_+ = \begin{pmatrix} 0 \\ B \end{pmatrix}, \quad C_+ = (0 \quad C)$$

Can use  $\text{place}()$  or  $\text{qr}()$  to determine  $K$ 's as before.

Note  $Q_+ = (Q_{+1} \quad Q_{+2} \quad Q_{+3} \quad Q_{+4})$  has four weights.

# Propeller Arm example



$$\frac{d}{dt} \begin{pmatrix} \alpha \\ \omega \\ \theta \end{pmatrix} = \begin{pmatrix} -\beta & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix} \begin{pmatrix} \alpha \\ \omega \\ \theta \end{pmatrix} + \begin{pmatrix} \gamma \\ 0 \\ 0 \end{pmatrix} u(t) \quad y(t) = (0 \quad 0 \quad 1) \begin{pmatrix} \alpha \\ \omega \\ \theta \end{pmatrix} \quad u(t) = K_r \theta_d - (K_1 \quad K_2 \quad K_3) \begin{pmatrix} \alpha \\ \omega \\ \theta \end{pmatrix}$$

$$x = \begin{pmatrix} \alpha \\ \omega \\ \theta \end{pmatrix} \quad A = \begin{pmatrix} -\beta & 0 & 0 \\ 1 & 0 & 0 \\ 0 & 1 & 0 \end{pmatrix} \quad B = \begin{pmatrix} \gamma \\ 0 \\ 0 \end{pmatrix} \quad C = (0 \quad 0 \quad 1) \quad K = (K_1 \quad K_2 \quad K_3)$$

# Example 1: Pole Placement

```
% Define all the matrices, using beta and gamma  
A(1,1) = -50;  
A(1,2) = 0.0;  
A(1,3) = 0.0;  
B(1) = 3000;  
.....
```

```
% Convert to a system data structure  
sys = ss(A, B, C, D, 'StateName', {s1,s2,s3});
```

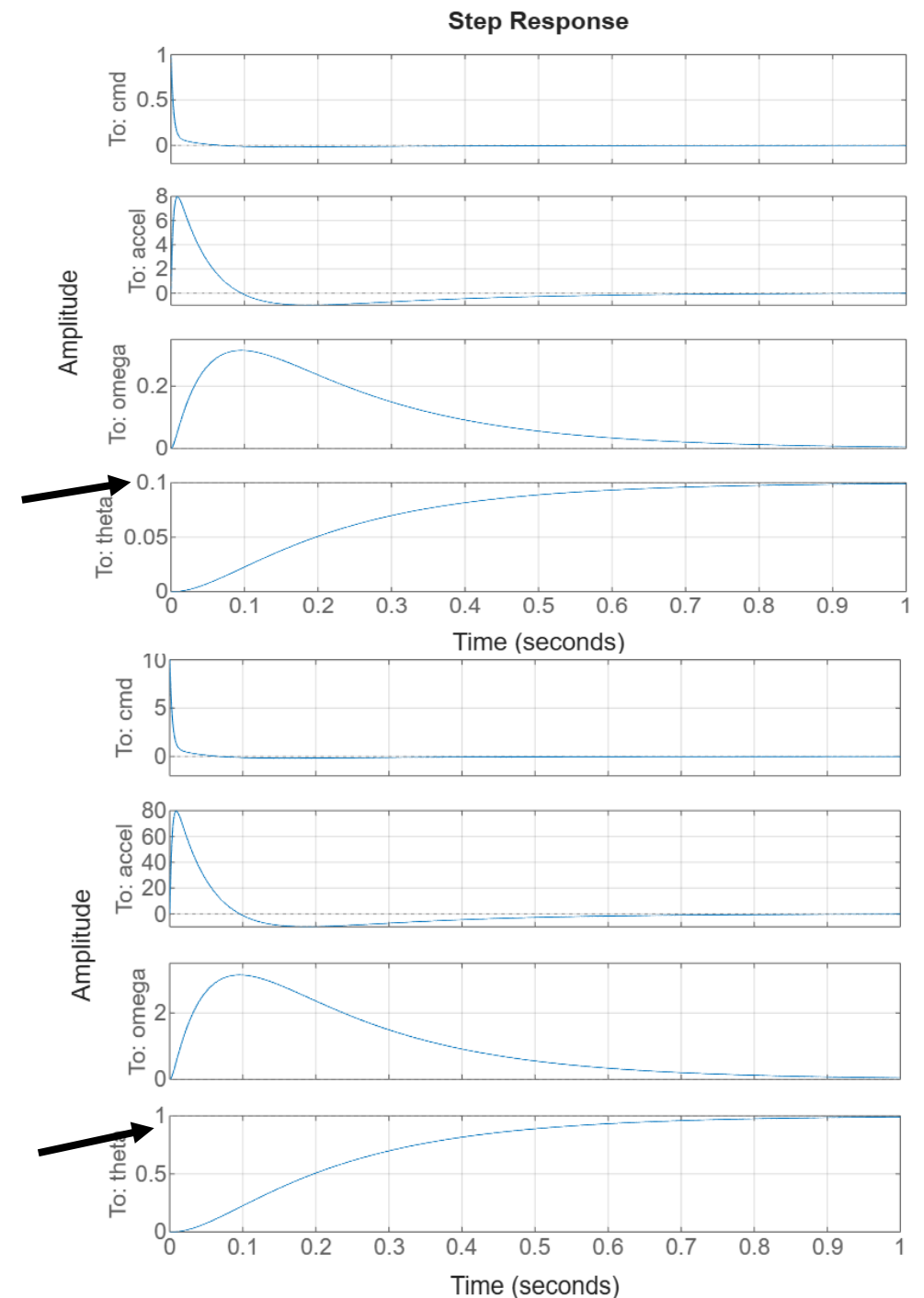
```
% Select closed loop poles for pole placement  
design
```

```
poles = [-5, -20, -300];  
kplace = place(sys.A,sys.B,poles);
```

```
Kr = 1;
```

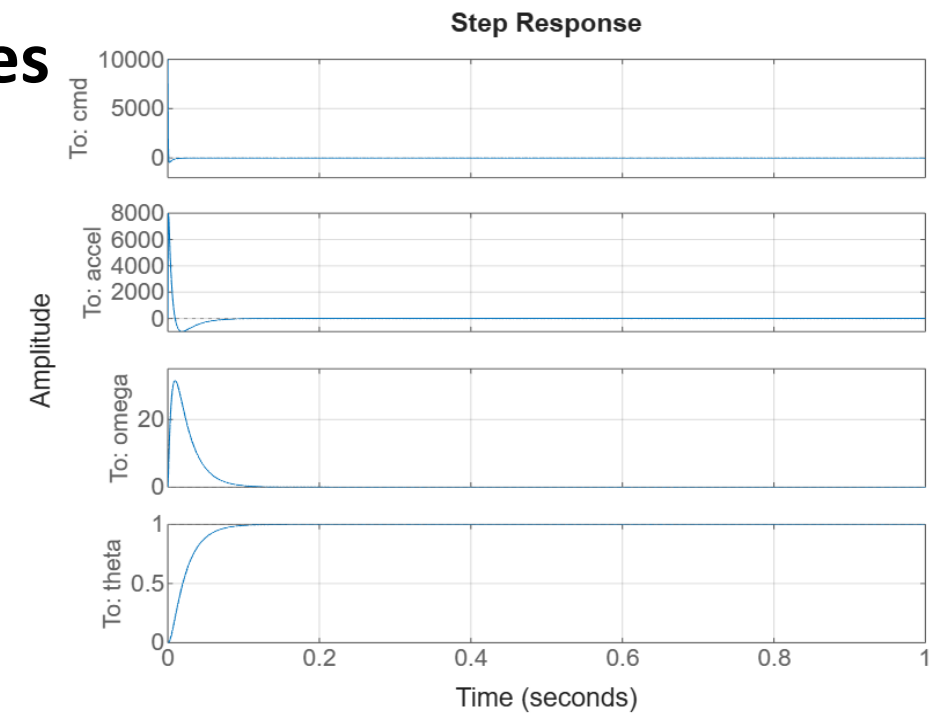
## Example 2: Fix Kr

```
% Use the Correct Kr  
Kr = -1/(C*(A-B*K)^(-1)*B);
```



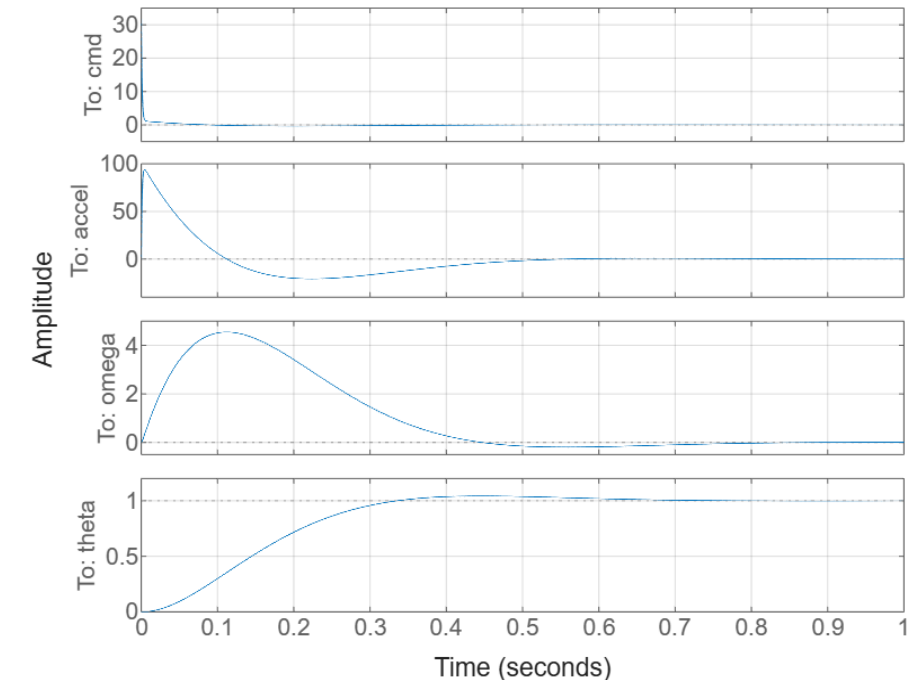
## Example 3: Place at a different set of poles

```
% Try to place the poles at different  
locations  
poles = [-50, -200, -3000];  
kplace = place(sys.A,sys.B,poles);
```



## Example 4: LQR

```
% Select state weights Q, and control weights R  
Q = diag([0.1,0.1,1000]);  
% Metric:  $Q(1,1)*x_1^2 + Q(2,2)*x_2^2 + Q(3,3)*x_3^2 + R*u^2$   
R = 1;  
Klqr = lqr(sys,Q,R);
```



## Example 5: Adding Integral Term

$$\frac{d}{dt} \begin{pmatrix} J \\ x \end{pmatrix} = \begin{pmatrix} 0 & C \\ 0 & A \end{pmatrix} \begin{pmatrix} J \\ x \end{pmatrix} + \begin{pmatrix} 0 \\ B \end{pmatrix} u(t) + \begin{pmatrix} -1 \\ 0 \end{pmatrix} y_d$$

$$y(t) = (0 \quad C) \begin{pmatrix} J \\ x \end{pmatrix} \quad u(t) = -(K_1 \quad K_2 \quad K_3 \quad K_4) \begin{pmatrix} J \\ \alpha \\ \omega \\ \theta \end{pmatrix}$$

%no  $K_r \theta_d$  , why?? (Recall your Postlab2)

```
% to add the integrator part to the
controller, we augment the original
% equations by the "integrator state" J
% dJ/dr = y - y_d .
```

```
.....
Ap = zeros(4);
Ap(2:4,:) = [zeros(3,1) A];
```

```
..... •
```

```
% define the new system with Ap, Bp, Cp and Dp
```

```
sysp = ss(Ap, Bp, Cp, Dp, 'StateName',{'Int',
s1, s2, s3});
```

```
% State weights
```

```
Qp = diag([1000000,0.1,0.1,1000]);
Rp = diag(1);
Kp = lqr(sysp,Qp,Rp);
```

